

Adaptive Finite-Time Control of Master-Slave Manipulators with Time-Varying Delay

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Adaptive finite-time control of master-slave manipulators with time-varying delay

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Abstract— The work presented in this paper focuses on the finite-time control of the master-slave manipulators with time-varying delay. First, a nonsingular fast terminal sliding mode control (NFTSMC) is used for this system based on the knowledge of the upper bound of the uncertainties and disturbances. Despite of the presence of the uncertainties, disturbances, load variation and time-varying delay, the controller is robust and present a finite-time convergence. Second, an adaptive nonsingular fast terminal sliding mode control (ANFTSMC) is applied to master-slave manipulators to avoid the knowledge of the upper bound of the uncertainties and disturbances. This controller presents good performance compared to the first one. The two controllers are evaluated in simulation.

Keywords—master-slave robot, non-singular fast terminal sliding mode control, adaptive control, uncertainties and disturbances, time-varying delay, robust control.

I. INTRODUCTION

Remote operation consists of an operator, a master side, a communication channel and a slave side that acts on an environment [1]. This system (master-slave robot) is a non-linear system with variable parameters and subject to external perturbations [2].

It is then necessary to build a robust and efficient control system with respect to the presence of uncertainties, external perturbations and communication delays.

In [10], the control method is based on the modeling of the teleoperation system in the state space, considering all the possible interaction in this system. The proposed controller is robust to the uncertainties, disturbances and time delay in the communication channel. In [9], the developed controllers take the advantage of the NFTSM theory to ensure fast convergence rate, singularity avoidance, and robustness against uncertainties and external disturbances. In [17], adaptive control achieves system robustness in the presence of uncertainties, perturbations and delays. In [18] [19], a sliding mode control is proposed to achieve the trajectory tracking performance based on exponential approach law and considering the variable payloads and model uncertainties.

Much research have used NFTSMC and its adaptive version that are applied on the robotic manipulators. However, there are no works of this type for the master-slave manipulators. In this paper, The NFTSM technique has some superior properties such as fast finite time convergence, high robustness and complete singularity avoidance. Therefore, the NFTSMC is applied to the teleoperation system in presence of the uncertainties, disturbances, load variation and time-varying delay. The controller needs the knowledge of the upper bound of the uncertainties and disturbances. In Neila Mezghani Ben Romdhane Laboratory of sciences and technique of automatic control and computing (Lab-STA) Nationnal school of engineering of sfax Sfax, Tunisia <u>neilamezghani@yahoo.fr</u>

practice, it is difficult to determine this bound. Then, an adaptive nonsingular fast terminal sliding mode control (ANFTSMC) is used to improve the performances of the NFTSMC and avoid the knowledge of the upper bound of the uncertainties and disturbances.

This work is organized as follows: In section 2, the system model is presented the non-singular fast terminal sliding mode control is described in section 3. In section 4, the adaptive version of the non-singular fast terminal sliding mode is exposed. Finally, some concluding remarks are exposed in section 5.

II. SYSTEM MODEL

The dynamics of the master-slave teleoperation can be defined as follows [14] [15] [16]:

$$M_{i}(q_{i})\ddot{q}_{i} + C_{i}(q_{i},\dot{q}_{i})\dot{q}_{i} + G_{i}(q_{i}) = u_{i} + P_{i}(t)$$
(1)

where $i \in \{m, s\}$. $q_i, \dot{q}_i, \ddot{q}_i \in \mathbb{R}^n$ are the position, velocity and acceleration signals for master and slave manipulators respectively. $M_i(q_i) \in \mathbb{R}^{n \times n}$ is the inertia matrices. $C_i(q_i, \dot{q}_i) \in \mathbb{R}^{n \times n}$ is coriolis/centrifugal force matrices. $G_i(q_i) \in \mathbb{R}^n$ is the gravitational force matrices. $P_i(t) \in \mathbb{R}^n$ is the external disturbance and modeling errors. n is the degree of freedom of manipulators. u_i is the control input torques of the teleoperations manipulators [6].

Assuming that the system described in (1) has known parts $M_{i0}(q_i), C_{i0}(q_i, \dot{q}_i) \quad G_{i0}(q_i)$ and unknown parts $\Delta M_i(q_i), \Delta C_i(q_i, \dot{q}_i)$ and $\Delta G_i(q_i)$, then

$$M_i(q_i) = M_{i0}(q_i) + \Delta M_i(q_i)$$
(2)

$$C_{i}\left(q_{i},\dot{q}_{i}\right) = C_{i0}\left(q_{i},\dot{q}_{i}\right) + \Delta C_{i}\left(q_{i},\dot{q}_{i}\right) \qquad (3)$$

$$G_{i}\left(q_{i}\right) = G_{i0}\left(q_{i}\right) + \Delta G_{i}\left(q_{i}\right) \tag{4}$$

From (2)-(4), (1) can be described in the following equation;

$$M_i(q_i)\ddot{q}_i + C_i(q_i,\dot{q}_i)\dot{q}_i + G_i(q_i) = u_i + F_i(q_i,\dot{q}_i,\ddot{q}_i)$$

Where

$$F_{i}(q_{i},\dot{q}_{i},\ddot{q}_{i}) = -\Delta M_{i}(q_{i})\ddot{q}_{i} - \Delta C_{i}(q_{i},\dot{q}_{i})\dot{q}_{i} - \Delta G_{i}(q_{i}) + P_{i}(t)$$
Assuming that [33]

Assuming that [33]

$$F_{i}(q_{i}, \dot{q}_{i}, \ddot{q}_{i}) = C_{i0} + C_{i1} |q_{i}| + C_{i2} ||\dot{q}_{i}||^{2}$$

The master-slave robot is a non-linear system with variable parameters, subject to external disturbances and communication delays at the slave. The problem is to design a robust and efficient control system with respect to the presence of uncertainties, external perturbations and communication delays.

III. NON-SINGULAR FAST TERMINAL SLIDING MODE CONTROL

A. Control design

The robot can be described by the following error system:

$$\begin{cases} \dot{e}_{i1} = e_{i2} \\ \dot{e}_{i2} = \ddot{q}_{id} - M_{i0}^{-1} (q_i) (-C_{i0}(q_i, \dot{q}_i) \dot{q}_i & i \in \{m, s\} \\ -G_{i0} (q_i) + u_i + F_i (q_i, \dot{q}_i, \ddot{q}_i)) \end{cases}$$
(5)

Where $e_{i1} = q_{id} - q_i$ and $e_{i2} = \dot{q}_{id} - \dot{q}_i$ $q_{md} = q_d$ for the master $q_{sd} = q_{md} (t - T)$ for the slave.

To apply the non-singular fast terminal sliding mode control to the master-slave manipulator, the first step is to choose the following sliding surface [4][5][8][7][9]:

$$s_{i} = e_{i1} + K_{i1} \left| e_{i1} \right|^{\beta_{i}} sign(e_{i1}) + K_{i2} \left| e_{i2} \right|^{\alpha_{i}} sign(e_{i2})$$
(6)

The control is determined using the equivalent control method with putting $\dot{s} = 0$ [12],

Then

$$\dot{e}_{i1} + \beta_i K_{i1} \ diag \ (\left|e_{i1}\right|^{\beta_i - 1}) \dot{e}_{i1} + \alpha_i K_{i2} \ diag \ (\left|\dot{e}_{i1}\right|^{\alpha_i - 1}) \dot{e}_{i2} = 0$$

Using (5), proposing that $F_i(q_i, \dot{q}_i, \ddot{q}_i) = 0$, we obtain:

$$\dot{e}_{i1} + \beta_i K_{i1} \ diag \ \left(\left|e_{i1}\right|^{\beta_i - 1}\right) \dot{e}_{i1} + \alpha_i K_{i2} \ diag \ \left(\left|\dot{e}_{i1}\right|^{\alpha_i - 1}\right)$$
$$\left(\ddot{q}_{id} - M_{i0}^{-1}\left(q_i\right)\left(-C_{i0}\left(q_i, \dot{q}_i\right)\dot{q}_i - G_{i0}\left(q_i\right) + u_{ieq}\right) = 0$$

Then, the equivalent control is

$$u_{ieq} = \frac{M_{i0}}{\alpha_i K_{i2}} diag(|\dot{e}_{i1}|^{2-\alpha_i}) sign(\dot{e}_{i1})$$

$$(1 + \beta_i k_{i1} diag(|e_{i1}|^{\beta_i - 1})) + M_{i0}(q_i) \ddot{q}_{id} \qquad (7)$$

$$+ C_{i0}(q_i, \dot{q}_i) \dot{q}_i + G_{i0}(q_i)$$

The control is the sum of equivalent control u_{ieq} and the discontinuous control u_{id} .

$$u_i = u_{ieq} + u_{id} \tag{8}$$

Where

$$u_{id}(t) = M_i (K_i * s_i + (C_{i0} + C_{i1} || q_i || + C_{i2} || \dot{q}_i ||^2) sign(s_i))$$
(9)

B. Stabilty Analysis

To verify the stability of the control, the following Lyapunov function is considered

$$\dot{V} = s_i^T \dot{s}_i \tag{10}$$

$$\dot{V} = s_i^T (\dot{e}_{i1} + \beta_i K_i \ diag \ (\left|e_{i1}\right|^{\beta_i - 1}) \dot{e}_{i1} + \alpha_i K_{i2} \ diag \ (\left|\dot{e}_{i1}\right|^{\alpha_i - 1}) \dot{e}_{i2})$$
(11)

Remplacing the control (8) in the expression of the derivative lyapunov, we have:

$$\dot{V} = s_i^T (\dot{e}_{i1} + \beta_i K_{i1} \ diag \ (\left|e_{i1}\right|^{\beta_i - 1}) \dot{e}_{i1} + \alpha_i K_{i2}$$

$$diag \ (\left|\dot{e}_{i1}\right|^{\alpha_i - 1}) \ (\ddot{q}_{id} - M_{i0}^{-1}(q_i) \ ($$

$$-C_{i0} \ (q_i, \dot{q}_i) \dot{q}_i - G_{i0} \ (q_i) + u_i + F_i \ (q_i, \dot{q}_i, \ddot{q}_i))))$$
(12)

Using equation (9), the derivative of V is

$$\dot{V} = -\alpha_i K_{i2} \ diag(\left|\dot{e}_{i1}\right|^{\alpha_i - 1}) (s_i^T M_{i0}^{-1} u_{id} - s_i^T F_i(q_i, \dot{q}_i, \ddot{q}_i))$$
(13)

As a result

$$\dot{V} = -\alpha_i K_{i2} \ diag \ (\dot{e}_{i1} |^{\alpha_i - 1}) (s_i^T K s_i + s_i^T (C_{i0} + C_{i1} \| q_i \| + C_{i2} \| \dot{q}_i \|^2) sign(s_i) - s_i^T F_i(q_i, \dot{q}_i, \ddot{q}_i))$$
(14)

$$\begin{aligned} &\alpha_{i}K_{i2}diag\;\left(\left|\dot{e}_{i1}\right|^{\alpha_{i}-1}\right)\left(-s_{i}^{T}K_{i}\;s_{i}-s_{i}^{T}\left(C_{i0}+C_{i1}\mid\left|q_{i}\right|\right|\right.\\ &+C_{i2}\mid\left|\dot{q}_{i}\right|\right|^{2})sign\left(s_{i}\right)+s_{i}^{T}\;F_{i}\left(q_{i},\dot{q}_{i},\ddot{q}_{i}\right)\right)\\ &\leq\alpha_{i}K_{i2}diag\;\left(\left|\dot{e}_{i1}\right|^{\alpha_{i}-1}\right)\left(-s_{i}^{T}K_{i}\;s_{i}+\left(-\left|s_{i}\right|\left(C_{i0}+C_{i1}\right)\right.\\ &\left|\left|q_{i}\right|\right|+C_{i2}\mid\left|\dot{q}_{i}\right|\right|^{2}\right)sign\left(s_{i}\right)+\left|s_{i}\right|\left|\left|F_{i}\left(q_{i},\dot{q}_{i},\ddot{q}_{i}\right)\right|\right)\right)\\ &\leq-\alpha_{i}K_{i2}diag\;\left(\left|\dot{e}_{i1}\right|^{\alpha_{i}-1}\right)-s_{i}^{T}K_{i}\;s_{i}\;\;\leq0\end{aligned}$$

Then, the system is asymptotically stable.

IV. ADAPTATIVE NON-SINGULAR FAST TERMINAL SLIDING MODE CONTROL

In practice, the upper bound of the system uncertainty is often unknown in advance and hence the components of the vector uncertainty $F_i(q_i, \dot{q}_i, \ddot{q}_i)$ are difficult to find. Then an adaptive tuning low is used to estimate C_{i0} , C_{i1} and C_{i2} . An adaptive non-singular fast terminal sliding mode control (ANFTSMC) has been studied to improve system output in the presence of parametric uncertainties, external disturbances and communication delay.

The same sliding surface (6) and equivalent control (7) are used.

The discontinuous control is [20]

$$u_{id}(t) = M_{i0} (K_i * s_i + (a_i + \hat{C}_{i0} + \hat{C}_{i1} || q_i || + \hat{C}_{i2} || \dot{q} ||^2) sign(s_i))$$
(16)

Where a_i is a positive constant.

To study the stability of this control, we consider the following Lyapunov function:

$$V = V_1 + V_2$$
(17)

$$V_1 = \frac{1}{2} s_i^2$$
(18)

$$V_{2} = \alpha K_{2} \sum_{i=0}^{2} \frac{1}{2\gamma_{i}} \tilde{C}_{ij}^{2}$$
(19)

Where γ_i are positive constants.

$$\tilde{C}_{ij} = \hat{C}_{ij} - C_{ij} \tag{20}$$

Where \hat{C}_i is the estimated value of C_i and \tilde{C}_i is the estimation error.

The derivative of (18) is

$$\begin{split} \dot{V_{1}} &= s_{i}^{T} \dot{s}_{i} = s_{i}^{T} (\dot{e}_{i1} + \beta_{i} K_{i1} \ diag \ (\left|e_{i1}\right|^{\beta_{i}-1}) \\ \dot{e}_{i1} + \alpha_{i} K_{i2} \ diag \ (\left|\dot{e}_{i1}\right|^{\alpha_{i}-1}) \dot{e}_{i2}) \\ &= s_{i}^{T} (\dot{e}_{i1} + \beta_{i} K_{i1} \ diag \ (\left|e_{i1}\right|^{\beta_{i}-1}) \dot{e}_{i1} \\ &+ \alpha_{i} K_{i2} \ diag \ (\left|\dot{e}_{i1}\right|^{\alpha_{i}-1}) \ (\ddot{q}_{id} - M_{i0}^{-1}(q_{i}) \\ (-C_{i0}(q_{i}, \dot{q}_{i}) \dot{q}_{i} - G_{i0}(q_{i}) + u_{i} + F_{i}(q_{i}, \dot{q}_{i}, \ddot{q}_{i})))) \end{split}$$

For the system to be stable, it is necessary that $\dot{V}_1 < 0$ then the discontinuous control be as the following:

$$u_{id}(t) = M_{i0} (K_{i}s_{i} + (a_{i} + \hat{C}_{i0} + \hat{C}_{i0} || q_{i} || + \hat{C}_{i2} || \dot{q}_{i} ||^{2}) sign(s_{i}))$$

$$\dot{V}_{1} = \alpha_{i}K_{i2} diag(|\dot{e}_{i1}|^{\alpha_{i}-1})[-s_{i}^{T}K_{i}s_{i} - s_{i}^{T}(a_{i} + \hat{C}_{i0} + \hat{C}_{i1} || q_{i} || + \hat{C}_{i2} || \dot{q}_{i} ||) sign(s_{i}) + s_{i}^{T}F_{i}(q_{i}, \dot{q}_{i}, \ddot{q}_{i})]$$
(21)
$$(21)$$

The derivative of (19) is

$$\dot{V}_{2} = \alpha_{i} K_{i2} \sum_{i=0}^{2} \frac{1}{\gamma_{i}} = (\hat{C}_{ij} - C_{ij}) \dot{\hat{C}}_{ij}$$
(23)

Finally, the equations (22) and the control (7) (8) (16) in the equation (21) we obtain

$$\dot{V} = \dot{V_1} + \dot{V_2}$$

$$\dot{V} = \alpha_i K_{i2} \, diag \left(\left| \dot{e}_{i1} \right|^{\alpha_i - 1} \right) \left(-s_i^T K_i s_i - s_i^T (a_i + \hat{C}_{i0} + \hat{C}_{i1} \| q_i \| + \hat{C}_{i2} \| \dot{q}_i \| \right) sign(s_i) + s_i^T F_i(q_i, \dot{q}_i, \ddot{q}_i)$$

$$+ \alpha_i K_{i2} \sum_{j=0}^2 \frac{1}{\gamma_{ij}} (\hat{C}_{ij} - C_{ij}) \dot{C}_{ij}$$

$$(25)$$

So that to have $\dot{V} < 0$ you need [11] [13] [20]

$$\dot{\hat{C}}_{i0} = \gamma_{i0} s_i^T \operatorname{diag}\left(\left|\dot{e}_i\right|^{\alpha_i - 1}\right) \operatorname{sign}\left(s_i\right)$$
$$\dot{\hat{C}}_{i1} = \gamma_{i1} s_i^T \operatorname{diag}\left(\left|\dot{e}_{i1}\right|^{\alpha_i - 1}\right) \operatorname{sign}\left(s_i\right) \parallel q_i \parallel \qquad (26)$$
$$\dot{\hat{C}}_{i2} = \gamma_{i2} s_i^T \operatorname{diag}\left(\left|\dot{e}_{i1}\right|^{\alpha_i - 1}\right) \operatorname{sign}\left(s_i\right) \parallel q_i \parallel^2$$

$$\dot{V} = \alpha_i K_{i2} \ diag\left(\left|\dot{e}_{i1}\right|^{\alpha_i - 1}\right) \left(-s_i^T K_i s_i - s_i^T (a_i + C_{i0} + C_{i1} \|q_i\| + C_{i2} \|q_i\|^2) sign(s_i) + s_i^T F_i(q_i, \dot{q}_i, \ddot{q}_i)\right)$$

Replace (25) in (24), we obtain [20]

$$\begin{split} \dot{V} &= \alpha_{i}K_{i2} \ diag\Big(\Big|\dot{e}_{i1}\Big|^{\alpha_{i}-1}\Big)(-s_{i}^{T}K_{i}s_{i} - s_{i}^{T}(a_{i} + \hat{C}_{i0} \\ &+ \hat{C}_{i1} \parallel q_{i} \parallel + \hat{C}_{i2} \parallel q_{i} \parallel^{2})sign(s_{i})) + s_{i}^{T}F_{i}(q_{i}, \dot{q}_{i}, \ddot{q}_{i}) \\ &+ \alpha_{i}K_{i2}((\hat{C}_{i0} - C_{i0})s_{i}^{T}diag \mid \dot{e}_{i1} \mid^{\alpha_{i}-1}sign(s_{i}) \\ &+ (\hat{C}_{i1} - C_{i1})s_{i}^{T}diag \mid \dot{e}_{i1} \mid^{\alpha_{i}-1}sign(s_{i}) \parallel q_{i} \parallel \\ &+ (\hat{C}_{i2} - C_{i2})s_{i}^{T}diag \mid \dot{e}_{i1} \mid^{\alpha_{i}-1}sign(s_{i}) \parallel q_{i} \parallel^{2}) \end{split}$$

After simplification, we get

$$\dot{V} = -\alpha_{i}K_{i2}s_{i}^{T}diag\left(\left|\dot{e}_{i1}\right|^{\alpha_{i}-1}\right)\left(s_{i}^{T}K_{i}s_{i}^{T}-s_{i}^{T}\left(a_{i}^{T}+C_{i0}^{T}+C_{i1}^{T}\right)\right) + C_{i1}\left\|q_{i}^{T}\right\| + C_{i2}\left\|q_{i}^{T}\right\|^{2}sign\left(s_{i}^{T}\right)\right) + s_{i}^{T}F_{i}\left(q_{i}^{T},\dot{q}_{i}^{T},\ddot{q}_{i}^{T}\right)\right) \\ \dot{V} \leq 0 \forall \left|\dot{e}_{i}^{T}\right| \neq 0$$
(27)

V. SIMULATION RESULT

In order to evaluate the effectiveness of the proposed controller, it has been applied on two degree-of-freedom teleoperations systems. The dynamic description of the master-slave manipulator is given as in the following [6].

$$M_i\left(q_i\right)\ddot{q}_i + C_i\left(q_i, \dot{q}_i\right)\dot{q}_i + G_i\left(q_i\right) = u_i + P_i(t)$$
$$M_i\left(q_i\right) = \begin{pmatrix} M_{11} & M_{12} \\ M_{21} & M_{22} \end{pmatrix}$$

$$C_{i}\left(q_{i}, \dot{q}_{i}\right) = \begin{pmatrix} C_{11} & C_{12} \\ C_{21} & C_{22} \end{pmatrix}$$
$$G_{i}\left(q_{i}\right) = \begin{pmatrix} G_{1} \\ G_{2} \end{pmatrix}$$

With

$$\begin{split} M_{11} &= m_{i2}l_{i2}^{2} + l_{i2}^{2}\left(m_{i1} + m_{i2}\right) + 2l_{i1}l_{i2}m_{i2}\cos\left(q_{i2}\right) \\ M_{12} &= M_{12} = m_{i2}l_{i2}^{2} + l_{i1}l_{i2}m_{i2}\cos\left(q_{i2}\right) \\ M_{22} &= m_{i2}l_{i2}^{2} \\ C_{11} &= -2l_{i1}l_{i2}m_{i2}\sin\left(q_{i2}\right)\dot{q}_{i2} \\ C_{12} &= -l_{i1}l_{i2}m_{i2}\sin\left(q_{i2}\right)\dot{q}_{i2} \\ C_{21} &= l_{i1}l_{i2}m_{i2}\sin\left(q_{2}\right)\dot{q}_{i1} \\ C_{22} &= 0 \end{split}$$

$$G_2 = g_i l_{i2} m_{i2} cos(q_{i2} + q_{i1})$$

The parameters of the system are $m_{m2} = m_{s2} = 0.5 Kg \pm 10\%$,

$$m_{m1} = m_{s1} = 4Kg \pm 10\%, \ l_{m1} = l_{s1} = l_{m2} = l_{s2} = 0.5m$$

and $g_i = 9.81NKg^{-1}$, initials state are $\begin{bmatrix} q_{m1} \\ q_{m2} \end{bmatrix} = \begin{bmatrix} 0.6 \\ 0.1 \end{bmatrix}$ and $\begin{bmatrix} q_{s1} \\ q_{s2} \end{bmatrix}$
 $= \begin{bmatrix} -4 \\ -5 \end{bmatrix}$, the desired joints are $\begin{bmatrix} q_{m1d} \\ q_{m2d} \end{bmatrix} = \begin{bmatrix} \cos(t) \\ \cos(t) \end{bmatrix}$ and $\begin{bmatrix} q_{s1d} \\ q_{s2d} \end{bmatrix}$
 $= \begin{bmatrix} q_{m1} \\ q_{m2} \end{bmatrix}, \ P_m = P_s = \sin(t)$ and the delay is:
 $T = 1 - 0.5e^{-0.5t}$.

A. Non-Singular Fast Terminal Sliding Mode Control

The Simulink block of the non-singular fast terminal sliding mode control NFTSMC of the master-slave manipulator with time varying delay and sudden load variation is considered.



Fig. 1. Simulink block of the NFTSMC of the master-slave manipulator with time varying delay and sudden load variation





Fig. 3. Tracking of the first joint of the master by the NFTSMC

 $G_{1} = g_{i} l_{i2} m_{i2} \cos \left(q_{i2} + q_{i1} \right) + l_{i1} \left(m_{i1} + m_{i2} \right) \cos \left(q_{i1} \right)$

Fig. 2. Mass uncertainties



Fig. 4. Tracking of the second joint of the master by the NFTSMC



Fig. 5. Tracking of the first joint of the slave by the NFTSMC



Fig. 6. Tracking of the second joint of the slave by the NFTSMC

For $\alpha_m = \alpha_s = 1.5$, $\beta_m = \beta_s = 2$, $K_{m1} = K_{s1} = 1$, $K_{m2} = K_{s2} = 1$, $C_{m0} = C_{s0} = 9.5$, $C_{m1} = C_{s1} = 2.2$, $C_{m2} = C_{s2} = 2.8$, a variable delay, equal to $T = 1 - 0.5e^{-0.5t}$ is added, and the robot carries a load equal to 100 g at time t = 5s. The simulation results Figs 3-6 show that the error tends towards zero and the sliding surface remains around zero. The trajectory in the phase plan slides in the vicinity of the sliding surface until it reaches the origin. The angular position figures present that the slave follows the master but with a shift due to the variable delay, then this method gives unsatisfactory results for the presence of varying

delays. In the control figures, a high frequency switching is noted, that is the problem of chattering.

In spite of the presence of disturbances, uncertainties and variable delays, the system always remains stable so that it converges rapidly towards a sliding surface.

B. Adaptive Non-Singular Fast Terminal Sliding Mode Control

The same Simulink block of NFTSMC of the masterslave manipulator with the time varying delay and sudden load variation is considered.



Fig. 7.Tracking of the first joint of the master by the ANFTSMC



Fig. 8. Tracking of the second joint of the master by the ANFTSMC



Fig. 9. Tracking of the first joint of the slave by the ANFTSMC



Fig. 10. Tracking of the second joint of the slave by the ANFTSMC

For the same values of the system parameters, a variable delay equal to $T=1-0.5e^{-0.5t}$, and that the robot carries a load equal to 100 g at the instant t=5s.

The Figs 7-10 of the angular position illustrate that the slave follows the master with a very reduced response time despite the presence of delay (the presence of delay does not affect the output). The rate of the tracking error shows the convergence of the system towards zero in the order of one second.

It can be seen that the sliding surface remains around zero, and the trajectory in the phase plan slides in the vicinity of the sliding surface until it reaches the origin.

This control eliminates the influence of the varying time delay on the slave.

VI. CONLUSION

In this paper, two controllers are guaranteeing a finitetime convergence and robustness to the uncertainties, disturbances, load variation and time-varying delay for the master-slave manipulators are used. The non-singular fast terminal sliding mode control is used in the case of the knowledge of the upper bound of the uncertainties and disturbances. In order to improve this controller, an adaptive nonsingular fast terminal sliding mode control is tested to the master-slave manipulator by avoiding the knowledge of the upper bound of the uncertainties and disturbances. Simulation results showed its better performance. In our future work, we will concentrate with the adaptive second-order control on bilateral teleoperations.

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